

These relations between the even/odd mode characteristic impedances and the differential/common mode characteristic impedances are consistent with the matched load terminations discussed in the literature [7], [8].

Now that voltages, currents, and characteristic impedances have been defined for both differential and common modes, the normalized power waves can be developed. By the definition for a generalized power wave at the  $n$ th port [23], [24]

$$\begin{aligned} a_n &= \frac{1}{2\sqrt{\text{Re}(Z_n)}} [v_n + i_n Z_n] \\ b_n &= \frac{1}{2\sqrt{\text{Re}(Z_n)}} [v_n - i_n Z_n^*] \end{aligned} \quad (16)$$

where  $a_n$  is the normalized wave propagating in the forward (positive  $x$ ) direction,  $b_n$  is the normalized wave propagating in reverse (negative  $x$ ) direction, and  $Z_n$  is the characteristic impedance of the port. With the above definitions, the differential normalized waves become, at port 1

$$\begin{aligned} a_{dm1} &\equiv a_{dm}(0) = \frac{1}{2\sqrt{\text{Re}(Z_{dm})}} [v_{dm}(x) + i_{dm}(x)Z_{dm}]|_{x=0} \\ b_{dm1} &\equiv b_{dm}(0) = \frac{1}{2\sqrt{\text{Re}(Z_{dm})}} [v_{dm}(x) - i_{dm}(x)Z_{dm}^*]|_{x=0}. \end{aligned} \quad (17)$$

Similarly, define the common-mode normalized waves, at port 1, as

$$\begin{aligned} a_{cm1} &\equiv a_{cm}(0) = \frac{1}{2\sqrt{\text{Re}(Z_{cm})}} [v_{cm}(x) + i_{cm}(x)Z_{cm}]|_{x=0} \\ b_{cm1} &\equiv b_{cm}(0) = \frac{1}{2\sqrt{\text{Re}(Z_{cm})}} [v_{cm}(x) - i_{cm}(x)Z_{cm}^*]|_{x=0}. \end{aligned} \quad (18)$$

Analogous definitions at port 2 can easily be found by setting  $x = l$ .

Imposing the condition of low-loss transmission lines on the coupled-pair of Fig. 2, the characteristic impedances are approximately purely real [24]. Under this restriction,  $Z_{dm} \approx \text{Re}\{Z_{dm}\} \equiv R_{dm}$  and  $Z_{cm} \approx \text{Re}\{Z_{cm}\} \equiv R_{cm}$ . With this assumption, the normalized wave equations at port 1 can be simplified

$$\begin{aligned} a_{dm1} &= \frac{1}{2\sqrt{R_{dm}}} [v_{dm}(x) + i_{dm}(x)R_{dm}]|_{x=0} \\ b_{dm1} &= \frac{1}{2\sqrt{R_{dm}}} [v_{dm}(x) - i_{dm}(x)R_{dm}]|_{x=0} \end{aligned} \quad (19)$$

$$\begin{aligned} a_{cm1} &= \frac{1}{2\sqrt{R_{cm}}} [v_{cm}(x) + i_{cm}(x)R_{cm}]|_{x=0} \\ b_{cm1} &= \frac{1}{2\sqrt{R_{cm}}} [v_{cm}(x) - i_{cm}(x)R_{cm}]|_{x=0}. \end{aligned} \quad (20)$$

With the normalized power waves defined, the development of mixed-mode  $s$ -parameters is straight forward. The definition of generalized  $s$ -parameters [23], [24] is

$$[b] = [S][a] \quad (21)$$

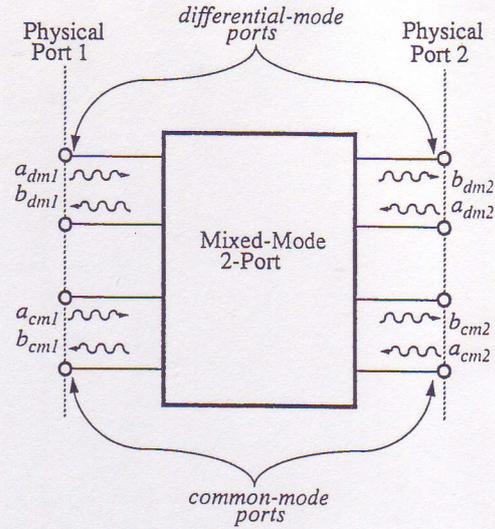


Fig. 3. Conceptual diagram of mixed-mode two-port.

where the bold letters denote an  $n$ -dimensional column vector or an  $n$ -by- $n$  matrix. Given a coupled-line two-port like Fig. 2, or any arbitrary mixed-mode two-port, the generalized mixed-mode  $s$ -parameters can be given as

$$\begin{aligned} b_{dm1} &= s_{11}a_{dm1} + s_{12}a_{dm2} + s_{13}a_{cm1} + s_{14}a_{cm2} \\ b_{dm2} &= s_{21}a_{dm1} + s_{22}a_{dm2} + s_{23}a_{cm1} + s_{24}a_{cm2} \\ b_{cm1} &= s_{31}a_{dm1} + s_{32}a_{dm2} + s_{33}a_{cm1} + s_{34}a_{cm2} \\ b_{cm2} &= s_{41}a_{dm1} + s_{42}a_{dm2} + s_{43}a_{cm1} + s_{44}a_{cm2} \end{aligned} \quad (22)$$

where the subscripts 1 and 2 denote ports 1 and 2, respectively. Here,  $[S]$  can be described by

$$\begin{bmatrix} b_{dm1} \\ b_{dm2} \\ b_{cm1} \\ b_{cm2} \end{bmatrix} = \begin{bmatrix} S_{dd} & S_{dc} \\ S_{cd} & S_{cc} \end{bmatrix} \begin{bmatrix} a_{dm1} \\ a_{dm2} \\ a_{cm1} \\ a_{cm2} \end{bmatrix} \quad (23)$$

The following names are used: Call  $[S_{dd}]$  the differential  $s$ -parameters,  $[S_{cc}]$  the common-mode  $s$ -parameters, and  $[S_{dc}]$  and  $[S_{cd}]$  the mode-conversion or cross-mode  $s$ -parameters. In particular,  $[S_{dc}]$  describes the conversion of common-mode waves into differential-mode waves, and  $[S_{cd}]$  describes the conversion of differential-mode waves into common-mode waves. These four partitions are analogues to four transfer gains ( $A_{cc}$ ,  $A_{dd}$ ,  $A_{cd}$ ,  $A_{dc}$ ) introduced by Middlebrook [2].

These mixed-mode two-port  $s$ -parameters can be shown graphically (see Fig. 3) as a traditional four-port. It must be remembered, however, that the ports are conceptual tools only, and not physically separate ports.

#### IV. CONSIDERATIONS FOR A PRACTICAL MIXED-MODE MEASUREMENT SYSTEM

The most straightforward means of implementing a mixed-mode  $s$ -parameter measurement system is to directly apply differential and common-mode waves while measuring the resulting differential and common-mode waves. Unfor-